Linear Single Cell Li-Ion Battery Charger with Auto Power Path Management

General Description

The RT9519B is an integrated single cell Li-ion battery charger with Auto Power Path Management (APPM). No external MOSFETs are required. The RT9519B enters sleep mode when power is removed. Charging tasks are optimized by using a control algorithm to vary the charge rate, including pre-charge mode, fast charge mode and constant voltage mode. For the RT9519B, the charge current can also be programmed with an external resistor and modified with an external GPIO. The scope that the battery regulation voltage can be modified with an external GPIO depends on the battery temperature. The internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures. The charging task will always be terminated in constant voltage mode when the charging current reduces to the termination current of 10% x I_{CHG FAST}. Other features include under voltage protection and over voltage protection for VIN the supply.

Ordering Information

RT9519B

Package Type QW : WQFN-20L 3x3 (W-Type) —Lead Plating System G : Green (Halogen Free and Pb Free)

Note :

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

Marking Information

46=YM DNN 46= : Product Code YMDNN : Date Code

Features

- 28V Maximum Rating for VIN Power
- Selectable Power Current Limit (0.1A / 0.5A / 1.5A)
- Integrated Power MOSFETs
- Auto Power Path Management (APPM)
- Battery Charging Current Control
- Battery Regulation Voltage Control
- Programmable Charging Current and Safe Charge Timer
- Under Voltage Protection, Over Voltage Protection
- Power Good and Charge Status Indicator
- Optimized Charge Rate via Thermal Feedback
- Thin 20-Lead WQFN Package
- RoHS Compliant and Halogen Free

Applications

- Digital Cameras
- PDAs and Smart Phones
- Portable Instruments

Pin Configurations



WQFN-20L 3x3



Typical Application Circuit



Function Block Diagram



Functional Pin Description

Pin No.	Pin Name	Pin Function
1	VIN	Supply Voltage Input.
2	PGOOD	Power Good Status Output. Open-drain output.
3	CHG	Charger Status Output. Open-drain output.
4, 21 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.
5	ISETA	Charge Current Set Input. Connect a resistor (RISETA) between ISETA and GND.
6	TS	Temperature Sense Input. The TS pin connects to a battery's thermistor to determine if the battery is too hot or too cold to charge. If the battery's temperature is out of range, charging is paused until it re-enters the valid range. TS also detects whether the battery (with NTC) is present or not.
7	TIMER	Safe Charge Timer Setting.
8,9	BAT	Battery Charge Current Output.
10	ĒN	Charge Enable. Active Low input. 200k Ω pull low.
11	ISET	Half Charge Current Set Input. Control by external GPIO, L = $I_{CHG1} / 2$, H = I_{CHG1} , 200k Ω pull low.
12	VP	This pin must be provided a regulated voltage from 2.95V to 3.6V by external power.
13	ISETU	VIN Current Limit Control Input. When ISETL = L, L = 100mA, H = 500mA, 200k Ω pull low.
14	ISETL	VIN Current Limit Control Input. L : see ISETU, H = 1.5A, 200k Ω pull low.
15	VSET	Battery Regulation Set Input. Control by external GPIO. L = 4.05V, H = 4.2V, 200k Ω pull low.
16	TS_FAULT	TS Fault Status Output. Open-drain Output. If TS is of fault (too cold, too hot) or VP triggers UVLO, $\overline{TS}FAULT = L$. If TS and VP are normal, $\overline{TS}FAULT = H$.
17, 18, 19	SYS	System Connect Pin. Connect this pin to System with a minimum $10\mu F$ ceramic capacitor to GND.
20	USUS	VIN Suspend Control Input. H = Suspend, L = No suspend. 200k Ω pull low.



Absolute Maximum Ratings (Note 1)

• Supply Voltage, V _{IN}	
• CHG, PGOOD	–0.3V to 28V
Other Pins	–0.3V to 6V
CHG, PGOOD Continuous Current	20mA
BAT Continuous Current (total in two pins) (Note 2)	2.5A
• Power Dissipation, $P_D @ T_A = 25^{\circ}C$	
WQFN-20L 3x3	1.471W
Package Thermal Resistance (Note 3)	
WQFN-20L 3x3, θ _{JA}	68°C/W
WQFN-20L 3x3, θ _{JC}	- 7.5°C/W
,	- 7.5°C/W
WQFN-20L 3x3, θ _{JC}	- 7.5°C/W 260°C
WQFN-20L 3x3, θ _{JC}	- 7.5°C/W 260°C 150°C
WQFN-20L 3x3, θ _{JC} • Lead Temperature (Soldering, 10 sec.) • Junction Temperature	- 7.5°C/W 260°C 150°C
 WQFN-20L 3x3, θ_{JC} Lead Temperature (Soldering, 10 sec.) Junction Temperature Storage Temperature Range 	- 7.5°C/W 260°C 150°C –65°C to 150°C

Recommended Operating Conditions (Note 5)

• Supply Input Voltage Range, VIN (ISETL = H)	- 4.4V to 6V
• Supply Input Voltage Range, VIN (ISETL = L)	- 4.5V to 6V
Junction Temperature Range	40°C to 125°C
Ambient Temperature Range	40°C to 85°C

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Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Supply Input						
VIN Under Voltage Lockout Threshold	V _{UVLO}	$V_{IN} = 0V \text{ to } 4V$	3.1	3.3	3.5	V
VIN Under Voltage Lockout Hysteresis	Δνυνίο	$V_{IN} = 4V$ to 0V		240		mV
		$I_{SYS} = I_{BAT} = 0mA, \overline{EN} = L (V_{BAT} > V_{REGx})$		1	2	
VIN Supply Current	ISUPPLY	$I_{SYS} = I_{BAT} = 0mA$, $\overline{EN} = H (V_{BAT} > V_{REGx})$		0.8	1.5	mA
VIN Suspend Current	lusus	$V_{IN} = 5V, USUS = H$		195	300	μA
VBAT Sleep Leakage Current	ISLEEP	$V_{BAT} > V_{IN}$, ($V_{IN} = 0V$)		5	15	μΑ
VIN-BAT VOS Rising	V _{OS_H}		-	200	300	mV
VIN-BAT VOS Falling	Vos_L		10	50		mV
Voltage Regulation						
System Regulation Voltage	Vsys	$I_{SYS} = 800 \text{mA}$	4.3	4.4	4.5	V
Battery Regulation Voltage	V _{REG1}	0 to 85°C , Loading = 20mA, When VSET = H	4.16	4.2	4.23	V
Battery Regulation Voltage	V _{REG2}	0 to 85°C, Loading = 20mA, When VSET = L	4.01	4.05	4.08	V
APPM Regulation Voltage	ΔV_{APPM}	$V_{SYS} - \Delta V_{APPM}$	120	200	280	mV
				-	To be co	ontinu

RT9519B

Parame	ter	Symbol	Test Conditions	Min	Тур	Max	Unit
DPM Regulation Volta	age	V _{DPM}	ISETL = L	4.3	4.4	4.5	V
VIN to VSYS MOSFE	T Ron	R _{DS(ON)}	I _{VIN} = 1000mA		0.2	0.35	Ω
BAT to VSYS MOSFE	ET Ron	R _{DS(ON)}	V _{BAT} = 4.2V, I _{SYS} = 1A		0.05	0.1	Ω
Re-Charge Threshold		ΔV_{REGCHG}	Battery Regulation – Recharge level	60	100	140	mV
Current Regulation							
ISETA Set Voltage (F Phase)	ast Charge	VISETA	$V_{BAT} = 4V, R_{ISETA} = 1k\Omega$		2		V
Charge Current Settin	ng Range	I _{CHG}		100		1200	mA
Charge Current Accur	racy1	I _{CHG1}	$V_{BAT} = 4V, R_{ISETA} = 1k\Omega$ ISET = H	570	600	630	mA
Charge Current Accur	racy2	I _{CHG2}	$V_{BAT} = 3.8V, R_{ISETA} = 1k\Omega$ ISET = L	285	300	315	mA
			ISETL = H (1.5A Mode)	1.2	1.5	1.8	Α
VIN Current Limit		I _{VIN}	ISETL = L, ISETU = H (500mA mode)	450	475	500	mA
			ISETL = L, ISETU = L (100mA Mode)	90	95	100	mA
Pre-charge							
BAT Pre-Charge Thre	shold	V _{PRECH}	BAT Falling	2.7	2.8	2.9	V
BAT Pre-Charge Thre	shold Hysteresis	Δ_{VPRECH}			200		mV
Pre-Charge Current		I _{CHG_PRE}	$V_{BAT} = 2V$	5	10	15	%
Charge Termination	Detection						
Termination Current F Charge (Except USB		I _{TERM}	ISETL = L, ISETU = H ISETL = H, ISETU = X	5	10	15	%
Termination Current Ratio to Fast Charge (USB100 Mode)		I _{TERM2}	ISETL = L, ISETU = L		3.3		%
Login Input/Output							
CHG Pull Down Volta	ge	V CHG	I _{CHG} = 5mA		200		mV
PGOOD Pull Down Vo	oltage	VPGOOD	I _{PGOOD} = 5mA		200		mV
TS_FAULT Pull Dowr	n Voltage	VTS_FAULT	I _{TS_FAULT} = 5mA		200		mV
EN, ISETL, USUS, ISETU, VSET, ISET	Logic-High	Vih		1.5			V
Threshold Voltage Logic-Low		V _{IL}				0.4	V
Protection							
Thermal Regulation		T _{REG}			125		°C
Thermal Shutdown Te	•	T _{SD}			155		°C
Thermal Shutdown Hy		ΔT _{SD}			20		°C
Over Voltage Protecti	on	Vovp	V _{IN} Rising	6.25	6.5	6.75	V
Over Voltage Protection	-	ΔV_{OVP}	V_{IN} = 7V to 5V, $V_{OVP} - \Delta V_{OVP}$		100		mV
Output Short Circuit D Threshold	Detection	VSHORT	VBAT – VSYS		300		mV

To be continued

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Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Time						
Pre-Charge Fault Time	t _{PCHG}	$C_{TIMER} = 1\mu F (1/8 \times t_{FCHG})$	1440	1800	2160	S
Fast charge Fault Time	t _{FCHG}	C _{TIMER} = 1µF	11520	14400	17280	S
PGOOD Deglitch Time	t _{PGOOD}	Time measured from VIN : $0 \rightarrow 5V$ 1µs rise-time to PGOOD = L		1		S
Input Over Voltage Blanking Time	t _{OVP}			50		μS
Pre-Charge to Fast-Charge Deglitch Time	t _{PF}			25		ms
Fast-Charge to Pre-Charge Deglitch Time	t _{FP}			25		ms
Termination Deglitch Time	t _{TERMI}			25		ms
Recharge Deglitch Time	t _{RECHG}			100		ms
Input Power Loss to SYS LDO Tum-Off Delay Time	t _{NO_IN}			25		ms
Pack Temperature Fault Detection Deglitch Time	t _{TS}			25		ms
Short Circuit Deglitch Time	t _{SHORT}			250		μS
Short Circuit Recovery Time	t _{SHORT-R}			64		ms
Other						
VP (External used only)	Vvp		2.95		3.6	V
VP Under Voltage Lockout Threshold		Falling Threshold		0.8		V
TS Battery Detect Threshold	V _{TS}		2.75	2.85	2.95	V
NTC						
Low Temperature Trip Point	V _{COLD}	Rising Threshold	73	74	75	% of VP
Low Temperature Trip point Hysteresis	ΔV _{COLD}			1		% of VP
High Temperature Trip Point	V _{HOT}	Falling Threshold	27	28	29	% of VP
High Temperature Trip Point Hysteresis	ΔV _{HOT}			1		% of VP

Note 1. Stresses listed as the above "Absolute Maximum Ratings" may cause permanent damage to the device. These are for stress ratings. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may remain possibility to affect device reliability.

Note 2. Guraranteed by design.

- Note 3. θ_{JA} is measured in natural convection at $T_A = 25^{\circ}C$ on a high effective thermal conductivity four-layer test board of JEDEC 51-7 thermal measurement standard. The measurement case position of θ_{JC} is on the exposed pad of the package.
- Note 4. Devices are ESD sensitive. Handling precaution is recommended.
- Note 5. The device is not guaranteed to function outside its operating conditions.

Typical Operating Characteristics













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Applications Information

The RT9519B is a fully integrated single-cell Li-ion battery charger ideal for portable applications. The internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures. Other features include under voltage protection and over voltage protection.

Pre-charge Mode

When the output voltage is lower than 2.8V, the charging current will be reduce to a fast-charge current ratio set by R_{ISETA} to protect the battery life time.

Fast-charge Mode

When the output voltage is higher than 3V, the charging current will be equal to the fast-charge current set by $\ensuremath{\mathsf{R}_{\mathsf{ISETA}}}.$

Constant Voltage Mode

When the output voltage is near 4.2V, and the charging current fall below the termination current, after a deglitch time check of 25ms, the charger will become disabled and \overline{CHG} will go from L to H.

Re-charge Mode

When the chip is in charge termination mode, the charging current will gradually go down to zero. However, once the voltage of the battery drops to below 4.1V, there will be a deglitch time of 100ms and then the charging current will resume again.

Charging Current Decision

The charge current can be set according to the following equations :

If $I_{SET} = H$ (for I_{CHG1}) $I_{CHG_FAST} = \frac{V_{ISETA}}{R_{ISETA}} \times 300$ If $I_{SET} = L$ (for I_{CHG2}) $I_{CHG_FAST} = \frac{V_{ISETA}}{R_{ISETA}} \times 150$ $I_{CHG_PRE} = 10\% \times I_{CHG_FAST}$

Time Fault

During the fast charge phase, several events may increase the charging time.

For example the system load current may have activated the APPM loop which reduces the available charging current. the device has entered thermal regulation because the IC junction temperature has exceeded T_{REG} .

During each of these events, if $3V < V_{BAT} < 4.1V$, the internal charging time is slowed down proportionately to the reduction in charging current. However, once the duration exceed the fault time, the CHG output will flash at approximately 2Hz to indicate a fault condition and the charge current will be reduced to about 1mA.

$$t_{FCHG_true} = t_{FCHG} \times \frac{2V}{V_{ISETA}}$$

$$t_{FCHG_true} : modified timer in fast$$

$$t_{FCHG} : original timer in fast charger$$

$$t_{FCHG} = 14400 \text{ sec} \times (\frac{C_{TIMER}}{1\mu F})$$

$$t_{PCHG} = \frac{t_{FCHG}}{8}$$

$$t_{PCHG} : timer in pre-charge$$

Time fault release methods :

- (1) Re-plug power
- (2) Toggle EN
- (3) Enter/exit suspend mode
- (4) Remove Battery
- (5) OVP

If ISET = H (for
$$I_{CHG1}$$
)

timer = t_{FCHG}

If ISET = L (for I_{CHG2})

timer = 2 x t_{FCHG}

Power Good

VIN Power Good ($\overline{PGOOD} = L$)

Input State	PGOOD Output
VIN < VUVLO	High Impedance
$V_{UVLO} < V_{IN} < V_{BAT} + V_{OS_H}$	High Impedance
$V_{BAT} + V_{OS_H} < V_{IN} < V_{OVP}$	Low Impedance
$V_{IN} > V_{OVP}$	High Impedance

(3)

Charge State Indicator

Charge State	CHG Output
Charging	Low
Charging Suspended by Thermal Loop	(for first charge cycle)
Safety Timers Expired	2Hz Flash
Charging Done	
Recharging after Termination	High Impedance
IC Disabled or no Valid Input Power	

Battery Pack Temperature Monitoring

The battery pack temperature monitoring function can be realized by connecting the TS pin to an external Negative Temperature Coefficient (NTC) thermistor to prevent over temperature condition. Charging is suspended when the voltage at the TS pin is out of normal operating range. The internal timer is then paused, but the value is maintained. When the TS pin voltage returns back to normal operating range, charging will resume and the safe charge timer will continue to count down from the point where it was suspended. Note that although charging is suspended due to the battery pack temperature fault, the \overline{CHG} pin will continue to remain low and indicate charging.





Too Cold Temperature

 $R_{COLD} = R_{NTC}$

Too Hot Temperature

 $R_{HOT} = R_{NTC}$

$$\frac{R2 + R_{COLD}}{R_{COLD} + R1 + R2} = 0.74$$
 (1)

$$\frac{R2 + R_{HOT}}{R_{HOT} + R1 + R2} = 0.28$$

From (1), (2) $R1 = \frac{R_{COLD} - R_{HOT}}{2.457}$ $R2 = 0.389 \times R1 - R_{HOT}$ If R2 < 0 $\frac{R_{COLD}}{R_{COLD} + R1} = 0.74$ From (3) $R1 = \frac{R_{COLD}}{0.74} - R_{COLD}$

Charge Enable

When \overline{EN} is low, the charger turns on. When \overline{EN} is high, the charger turns off. \overline{EN} is pulled low for initial condition.

VIN input Current Limit

ISETL	ISETU VIN Input Cu	
L	L	95mA
L	Н	475mA
Н	Х	1.5A

Suspend Mode

Set USUS = H, and the charge will enter Suspend Mode. In the Suspend Mode, \overline{CHG} is in high impedance and $I_{USUS(MAX)} < 300 \mu A$.

Power Switch

(2)

For the RT9519B, there are three power scenarios:

(1)When a battery and an external power supply (USB or adapter) are connected simultaneously :

If the system load requirements exceed that of the input current limit, the battery will be used to supplement the current to the load. However, if the system load requirements are less than that of the input current limit, the excess power from the external power supply will be used to charge the battery.

(2)When only the battery is connected to the system :

The battery provides the power to the system.

(3)When only an external power supply is connected to the system :

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The external power supply provides the power to the system.

Input DPM Mode

For the RT9519B, the input voltage is monitored when the USB100 or USB500 is selected. If the input voltage is lower than V_{DPM} , the input current limit will be reduced to stop the input voltage from dropping any further. This can prevent the IC from damaging improperly configured or inadequately designed USB sources.

APPM Mode

Once the sum of the charging and system load currents becomes higher than the maximum input current limit, the SYS pin voltage will be reduced. When the SYS pin voltage is reduced to the V_{APPM} , the RT9519B will automatically operate in APPM mode. In this mode, the charging current is reduced while the SYS current is increased to maintain system output. In APPM mode, the battery termination function is disabled.

Battery Supplement Mode Short Circuit Protect

In APPM mode, the SYS voltage will continue to drop if the charge current is zero and the system load increases beyond the input current limit. When the SYS voltage decreases below the battery voltage, the battery will kick in to supplement the system load until the SYS voltage rises above the battery voltage. While in supplement mode, there is no battery supplement current regulation. However, a built in short circuit protection feature is available to prevent any abnormal current situations. While the battery is supplementing the load, if the difference between the battery and SYS voltage becomes more than the short circuit threshold voltage, SYS will be disabled. After a short circuit recovery time, t_{SHORT_R} , the counter will be restarted. In supplement mode, the battery termination function is disabled. Note that for the battery supply mode exit condition, $V_{BAT} - V_{SYS} < 0V$.

Thermal Regulation and Thermal Shutdown

The RT9519B provides a thermal regulation loop function to monitor the device temperature. If the die temperature rises above the regulation temperature, T_{REG} , the charge current will automatically be reduced to lower the die temperature. However, in certain circumstances (such as high VIN, heavy system load, etc.) even with the thermal loop in place, the die temperature may still continue to increase. In this case, if the temperature rises above the thermal shutdown threshold, T_{SD} the internal switch between VIN and SYS will be turned off. The switch between the battery and SYS will remain on, however, to allow continuous battery power to the load. Once the die temperature decreases by ΔT_{SD} , the internal switch between VIN and SYS will be turned on again and the device returns to normal thermal regulation.





APPM Profile

1.5A Mode :



	I _{SYS}	V _{SYS}	I _{VIN}	I _{BAT}
T1, T7	0	SYS Regulation Voltage	CHG_MAX	CHG_MAX
T2, T6	< I _{VIN_OC} – CHG_MAX	SYS Regulation Voltage	I _{SYS} + CHG_MAX	CHG_MAX
T3, T5	> I _{VIN_OC} – CHG_MAX < I _{VIN_OCs}	Auto Charge Voltage Threshold	V _{IN_OC}	$V_{\text{IN}_\text{OC}} - I_{\text{SYS}}$
T4	> I _{VIN_OC}	V _{BAT} – I _{BAT} x R _{DS(ON)}	V _{IN_OC}	I _{SYS} -I _{VIN_OC}

USB 500mA Mode :



	I _{SYS}	V _{SYS}	I _{USB}	I _{BAT}
T1, T7	0	SYS Regulation Voltage	CHG_MAX	CHG_MAX
T2, T6	$< I_{VIN_{OC}}$ (USB) – CHG_MAX	SYS Regulation Voltage	I _{SYS} + CHG_MAX	CHG_MAX
T3, T5	> I _{VIN_OC} (USB) – CHG_MAX < I _{VIN_OC} (USB)	Auto Charge Voltage Threshold	I _{VIN_OC} (USB)	I _{VIN_OC} (USB) – I _{SYS}
T4	> I _{VIN_OC} (USB)	V _{BAT} – I _{BAT} x R _{DS(ON)}	I _{VIN_OC} (USB)	$I_{SYS} - I_{VIN_OC}$ (USB)



VSET vs V_{REG} , ISET vs I_{CHG}



For JEITA Battery Temperature Standard : CV regulation voltage will change at the following battery Temp ranges 0°C to 10°C and 45°C to 60°C

CC regulation current will change at the following battery Temp ranges 0°C to 10°C and 45°C to 60°C

	4.16 to 4.2 to 4.2	3V		
4.01 to 4.05 to 4.08V			4.01 to 4.0	5 to 4.08V
0°C	10°C	45°C		60°C
				Temperature +/- 2°C
	I _{CHG} +/- 5%			
0.5 x I _{CHO} +/- 5%			0.5 x I _{CHG} -	/- 5%
				Temperature +/- 2°C



RT9519B Operation State Digram for Charging



Operation State Digram for TS PIN





Thermal Considerations

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula :

 $\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = (\mathsf{T}_{\mathsf{J}(\mathsf{MAX})} - \mathsf{T}_{\mathsf{A}}) / \theta_{\mathsf{J}\mathsf{A}}$

where $T_{J(MAX)}$ is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction to ambient thermal resistance.

For recommended operating condition specifications of the RT9519B, the maximum junction temperature is 125°C and T_A is the ambient temperature. The junction to ambient thermal resistance, θ_{JA} , is layout dependent. For WQFN-20L 3x3 packages, the thermal resistance, θ_{JA} , is 68°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at T_A = 25°C can be calculated by the following formula :

 $P_{D(MAX)}$ = (125°C - 25°C) / (68°C/W) = 1.471W for WQFN-20L3x3 package

The maximum power dissipation depends on the operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance, θ_{JA} . For the RT9519B package, the derating curve in Figure 3 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.



Figure 3. Derating Curves for RT9519B Package

The RT9519B is a fully integrated low cost single cell Liion battery charger ideal for portable applications. Careful PCB layout is necessary. For best performance, place all peripheral components as close to the IC as possible.

A short connection is highly recommended. The following guidelines should be strictly followed when designing a

PCB layout for the RT9519B.

- Input and output capacitor should be placed close to the IC and connected to ground plane. The input trace on the PCB should be placed far away from sensitive devices and shielded by the ground.
- The GND and exposed pad should be connected to a strong ground plane for heat sinking and noise protection.
- The connection of RISETA should be isolated from other noisy traces. A short wire is recommended to prevent EMI and noise coupling.





Figure 4. PCB Layout Guide

Outline Dimension



Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions	n Millimeters	Dimensions In Inches		
	Min	Max	Min	Max	
A	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A3	0.175	0.250	0.007	0.010	
b	0.150	0.250	0.006	0.010	
D	2.900	3.100	0.114	0.122	
D2	1.650	1.750	0.065	0.069	
E	2.900	3.100	0.114	0.122	
E2	1.650	1.750	0.065	0.069	
е	0.400		0.016		
L	0.350	0.450	0.014	0.018	

W-Type 20L QFN 3x3 Package

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